

Basal ganglia oscillations: the role of delays and external excitatory nuclei

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Abstract—Basal ganglia are interconnected deep brain structures involved in movement generation. Their beta-band oscillations (13-30Hz) are known to be linked to Parkinson's disease motor symptoms. In this paper, we provide conditions under which these oscillations may occur, by explicitly considering the role of the pedunculopontine nucleus (PPN). We analyze the existence of equilibria in the associated firing-rate dynamics and study their stability by relying on a delayed MIMO frequency analysis. These results are illustrated with simulations that confirm numerically the analytic predictions of our two main theorems.

I. INTRODUCTION

Basal ganglia are deep brain structures involved in voluntary motor control as well as cognitive and motivational processes [16]. They have been studied extensively in connection with a variety of pathological observations such as Parkinson's disease [21]. Some evidence suggests that the advance of parkinsonism is highly correlated to the presence of abnormal oscillations in the beta band (13-30 Hz) within the basal ganglia [1]. There exist several hypotheses about the origin of these oscillations. Some of them emphasize the cortical [22] or striatal [13] origin of the phenomenon. But another popular assumption is that these oscillations may originate from the system composed of two excitatory-inhibitory basal ganglia nuclei: the subthalamic nucleus (STN), which is an excitatory nucleus, and the globus pallidus pars externa (GPe), which is an inhibitory nucleus [19]. Since basal ganglia are highly interconnected with the pedunculopontine nucleus (PPN) [7], [12], [14], this nucleus might have an influence on their oscillatory activity. The aim of our paper is to quantify this influence, with the help of control theory tools.

To explore the origin of pathological oscillations in the basal ganglia, some works [20], [8] exploit a

microscopic approach in which every neuronal cell is modeled individually. Other works rely on firing-rate models [15], [10], [17], [18]. These models are formulated in terms of an ordinary differential equation that rules the evolution of the number of spikes per time-unit within the considered neuronal population [3]. In [15], the authors exploit a firing-rate model of the STN and GPe populations to derive analytical conditions under which beta oscillations occur. These conditions have been improved in [17], [18] to provide tighter conditions for the existence of oscillations. The particular role of PPN within the basal ganglia has been specifically addressed in [11], where the authors study how the PPN responds to physiological and pathological inputs of the basal ganglia.

Here we develop a mathematical model that describes the interaction between three neuronal populations: PPN, STN and GPe (see Section II). To analyze this model, we extend the approach developed in [17] for two nuclei only. We first study the existence and uniqueness of equilibrium points. We derive necessary and sufficient conditions for the existence of multiple equilibria. Additionally, we propose a sufficient condition for global asymptotic stability in the absence of delays (see Section III). Then, the system is linearized and the MIMO Nyquist stability criterion [2] is applied to the feedback loop of the linearized system. We derive conditions on the delays and interconnection gains for the asymptotic stability of the network, and hence the absence of pathological oscillations (see Section IV). These results are illustrated by numerical simulations (see Section V). The proofs can be found in the preprint version of our paper [4].

II. MODEL DESCRIPTION

As we mentioned in the introduction, our objective is to analyze pathological oscillations in the basal ganglia. To characterize the firing rate of neural populations in STN, GPe and PPN, we use the well described firing-rate model. The architecture of our model is shown in Figure 1. The STN neurons project excitatory axons to the GPe [16], while GPe neurons project inhibitory axons to the STN and to other GPe neurons [9]. The STN neurons project also excitatory axons to the PPN, which projects back excitatory axons to the STN [14]. Additionally, the

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STN and PPN nuclei receive inputs from cortex [12], [9], [14] and the GPe nucleus receives input from the striatum [9].

Like in similar models [15], [17], [18], the firing rates of the STN, GPe, and PPN populations, respectively, are ruled by the delayed differential equations

$$\begin{aligned}\tau_s \dot{x}_s &= S_s(c_s^p x_p(t - \delta_s^p) - c_s^g x_g(t - \delta_s^g) + u_s) - x_s \\ \tau_g \dot{x}_g &= S_g(c_g^s x_s(t - \delta_g^s) - c_g^g x_g(t - \delta_g^g) + u_g) - x_g \\ \tau_p \dot{x}_p &= S_p(c_p^s x_s(t - \delta_p^s) + u_p) - x_p\end{aligned}\quad (1)$$

where x_s , x_g , and x_p represent the firing rates of the STN, GPe and PPN neurons, respectively. By abuse of notation, we omit the dependency of \dot{x}_i and x_i on t , for $i \in \{s, g, p\}$. The positive gains c_s^p , c_p^s , c_g^s , c_s^g and c_g^g define the weight of the different synaptic interconnections between these three neuron populations. The variables u_s , u_g and u_p describe the external inputs, from the striatum and cortex, received by these populations. The time constants τ_s , τ_g and τ_p describe how rapidly the three populations react to the inputs. We assume that all the delays δ_s^p , δ_p^s , δ_s^g , δ_g^s and δ_g^g are nonnegative and constant. The scalar functions S_s , S_g and S_p define the activation functions of STN, GPe and PPN respectively.

Assumption 1: For each $i \in \{s, g, p\}$, the activation function $S_i : \mathbb{R} \rightarrow (0; 1)$ is continuously differentiable and strictly increasing. Its infimum is equal to 0 and its supremum is equal to 1. In addition, its derivative S_i' is upper-bounded and there exists at least one point at which it reaches its maximum denoted σ_i .

III. ANALYSIS IN THE ABSENCE OF DELAYS

A. Existence and uniqueness of equilibrium points

The system (1) is defined everywhere on \mathbb{R}^3 , since the activation functions are defined on \mathbb{R} . Nevertheless, one can check that the unit cube is invariant for this system.

Lemma 1: Under Assumption 1, for any constant inputs u_s , u_g and u_p , the unit cube

$$D := \{(x_s, x_g, x_p) \in \mathbb{R}^3 : x_s, x_g, x_p \in [0, 1]\}.$$

is positively invariant for the delayed system (1).

The following result analyzes the existence and multiplicity of equilibria of the dynamics (1).

Theorem 1: Under Assumption 1, we have that if

$$\sigma_p \sigma_s c_s^p c_p^s \leq 1 \quad (2)$$

then the system (1) has a unique equilibrium point, for each constant vector $(u_s^*, u_g^*, u_p^*) \in \mathbb{R}^3$. Otherwise, there exists a constant vector (u_s^*, u_g^*, u_p^*) for which the system (1) has at least three distinct equilibria. Moreover, if

$$\left(\sigma_p c_s^p c_p^s - \frac{1}{\sigma_s} \right) \left(c_g^g + \frac{1}{\sigma_g} \right) > c_s^g c_g^s \quad (3)$$

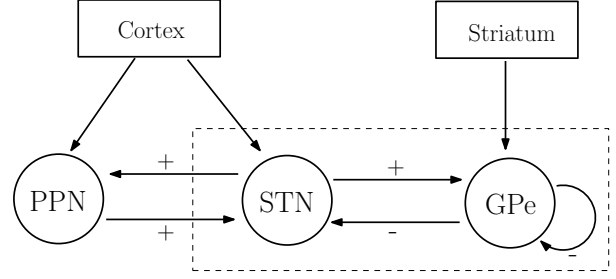


Fig. 1. Schematic diagram of three nuclei: STN, GPe and PPN.

then, for each constant input u_g^* , there exists a pair of constant input (u_s^*, u_p^*) for which the system (1) has at least three distinct equilibria.

Theorem 1 generalizes the equilibrium study given by [17, Theorem 1]. It shows, in particular, that under condition (3), the existence of multiple equilibria can occur even for arbitrarily small striatal input. The proof of Theorem 1 follows the same lines as in [17, Theorem 1] and can be found in [4].

B. Stability of equilibria

Consider an equilibrium point x^* , associated to a vector of inputs u^* , whose existence is ensured by Theorem 1. Let

$$e = x - x^* \quad \text{and} \quad v = u - u^*. \quad (4)$$

The linearization of the dynamics (1) around x^* is given by

$$\begin{aligned}\tau_s \dot{e}_s &= \sigma_s^* (c_s^p e_p(t - \delta_s^p) - c_s^g e_g(t - \delta_s^g) + v_s) - e_s \\ \tau_g \dot{e}_g &= \sigma_g^* (c_g^s e_s(t - \delta_g^s) - c_g^g e_g(t - \delta_g^g) + v_g) - e_g \\ \tau_p \dot{e}_p &= \sigma_p^* (c_p^s e_s(t - \delta_p^s) + v_p) - e_p,\end{aligned}\quad (5)$$

where

$$\begin{aligned}\sigma_s^* &:= S_s'(c_s^p x_p^* - c_s^g x_g^* + u_s^*) \\ \sigma_g^* &:= S_g'(c_g^s x_s^* - c_g^g x_g^* + u_g^*) \\ \sigma_p^* &:= S_p'(c_p^s x_s^* + u_p^*).\end{aligned}\quad (6)$$

We next rely on this linearization to study the stability properties of x^* . We start by considering system (1) in the absence of delays.

Proposition 1: Consider the delayed system (1), and assume that

$$\delta_i^j = 0 \quad \text{for} \quad i, j \in \{s, g, p\}. \quad (7)$$

Fix any input vector $u^* := (u_s^*, u_g^*, u_p^*)^T \in \mathbb{R}^3$, consider an equilibrium $x^* := (x_s^*, x_g^*, x_p^*)^T$ associated to these input, and let σ_i^* , $i \in \{s, g, p\}$ be defined by (6). Then, under Assumption 1, the following holds.

- If the conditions

$$\left(\sigma_p^* c_s^p c_p^s - \frac{1}{\sigma_s^*}\right) \left(c_g^g + \frac{1}{\sigma_g^*}\right) < c_s^g c_g^s \quad (8)$$

$$\frac{\sigma_s^*}{\tau_s + \tau_p} \left(\sigma_p^* c_s^p c_p^s - \frac{1}{\sigma_s^*}\right) < \frac{\sigma_g^*}{\tau_g} \left(c_g^g + \frac{1}{\sigma_g^*}\right) \quad (9)$$

are both satisfied, then the equilibrium point x^* is locally exponentially stable.

- If the conditions

$$\sigma_p \sigma_s c_s^p c_p^s < 1 \quad (10)$$

$$\sigma_s (c_s^p + c_s^g) + \sigma_g c_g^s + \sigma_p c_p^s < 2 \quad (11)$$

are both satisfied then x^* is globally asymptotically stable.

Section IV is devoted to the study of the preservation of these stability properties in the presence of delays.

IV. ROBUSTNESS TO DELAYS

In this section, we study the stability properties of the equilibria of (1) by relying on its linearization around an equilibrium point. This system can be described in the frequency domain using the closed-loop transfer functions

$$\begin{aligned} H_s(s) &= \frac{\sigma_s^*}{\tau_s s + 1}, & H_p(s) &= \frac{\sigma_p^*}{\tau_p s + 1} \\ H_g(s) &= \frac{\sigma_g^*}{\tau_g s + 1 + \sigma_g^* c_g^g e^{-\delta_g^g s}}, \end{aligned} \quad (12)$$

by the following feedback system (G, K)

$$\begin{cases} E = GE' + GV \\ E' = KE \end{cases} \quad (13)$$

where $E = (E_s, E_g, E_p)^T$, and $V = (V_s, V_g, V_p)^T$ are the Laplace transform of e and v (introduced in (4)), and the two transfer matrices G and K are given by the following

$$G(s) = \begin{pmatrix} H_s(s) & 0 & 0 \\ 0 & H_g(s) & 0 \\ 0 & 0 & H_p(s) \end{pmatrix}, \quad (14)$$

$$K(s) = \begin{pmatrix} 0 & -c_s^g e^{-\delta_s^g s} & c_s^p e^{-\delta_s^p s} \\ c_s^s e^{-\delta_s^s s} & 0 & 0 \\ c_p^s e^{-\delta_p^s s} & 0 & 0 \end{pmatrix}. \quad (15)$$

In order to study the stability properties of this linearized system, we make use of the Nyquist Theorem [2, Theorem 9.1.8] for MIMO delayed systems. We stress that here the Nyquist Theorem is applied in its general

form in the Callier-Desoer class of scalar irrational transfer functions $\hat{\mathcal{B}}$ (see [2, Definitions 7.1.4 and 7.1.6] for detailed definitions, see also [4, Appendix]). We start by checking the stability of the transfer matrices G and K . Noticing that G and K are irrational transfer matrices, we begin by verifying that each of its components belongs to $\hat{\mathcal{B}}$. This is stated by the following result, whose proof is provided in [4].

Proposition 2: The entries of the transfer matrices G and K defined in (14)-(15) all belong to the Callier-Desoer class of scalar irrational transfer functions $\hat{\mathcal{B}}$.

The transfer matrix G is not necessarily stable. This comes from the fact that the transfer function H_g is not necessarily stable. However, one can observe that H_g is always stable when $\delta_g^g = 0$. [17, Lemma 3] establishes the existence of a delay margin $\Delta(H_g)$ of H_g such that the transfer function H_g is input-output stable if and only if $\delta_g^g < \Delta(H_g)$. We recall that, for a $\bar{\tau} > 0$, the delay margin of an input-output stable SISO transmittance $H \in \hat{\mathcal{B}}$, is defined by

$$\begin{aligned} \Delta(H) &:= \sup\{\bar{\tau} > 0 : \text{the feedback } (H, e^{-\tau s}) \\ &\text{is input-output stable } \forall \tau \in [0, \bar{\tau})\}. \end{aligned}$$

A formal definition of input-output stability is given in [2, Definition 9.1.1], see also [4, Definition 1]. To compute the delay margin of H one can associate, at a given frequency ω , its gain $\gamma_H(\omega)$ and phase $\varphi_H(\omega)$ which are defined by the relations

$$\begin{aligned} \gamma_H(\omega) &= 20 \log_{10} |H(i\omega)| \\ \varphi_H(\omega) &= \arg(H(i\omega)) \end{aligned} \quad (16)$$

In our approach, the case where the function γ_H is strictly decreasing is of a particular interest. Indeed, in this case, to any strictly proper transfer function H such that $\gamma_H(0) > 0$ we can associate its (gain) crossover frequency ω_H , which is defined as the only frequency such that

$$\gamma_H(\omega_H) = 0. \quad (17)$$

This frequency can be used to define the delay margin $\Delta(H)$ by the relation

$$\Delta(H) = \frac{\pi - \varphi_H(\omega_H)}{\omega_H}. \quad (18)$$

If γ_H is strictly decreasing but $\gamma_H(0) \leq 0$, if H is minimum phase (it has neither unstable poles nor unstable zeroes), then we can still define $\Delta(H) = +\infty$.

Now that the stability properties of G and K are fully known in open loop, we analyze the stability of the feedback system (13). The following result provides a necessary and sufficient condition for the stability of the feedback system (13). Its statement relies on the

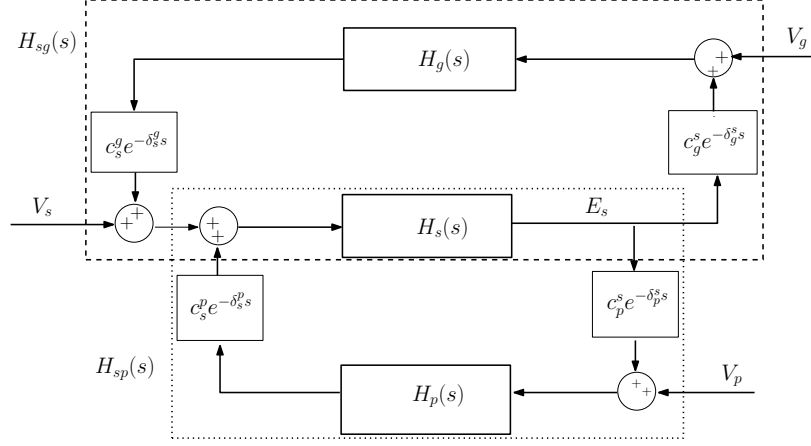


Fig. 2. Bloc diagram of the linearized system (5).

following two transfer functions

$$\begin{aligned} K_p(s) &:= c_p H_p(s) e^{-\delta_p s} \\ K_g(s) &:= -c_g H_g(s) e^{-\delta_g s} \end{aligned} \quad (19)$$

where the quantities

$$\begin{aligned} c_p &:= c_s^p c_p^s, & c_g &:= c_s^g c_g^s, \\ \delta_p &:= \delta_p^s + \delta_p^p, & \delta_g &:= \delta_g^s + \delta_g^g \end{aligned} \quad (20)$$

are defined in order to obtain a lighter notation.

Proposition 3: Suppose that H_g is input-output stable. The feedback system defined by (13) is input-output stable if and only if

$$\text{ind}(1 - H_s(K_p + K_g)) = 0, \quad (21)$$

where $\text{ind}(1 - H_s(K_p + K_g))$ denote the Nyquist index [2, Definition A.1.15] of $1 - H_s(K_p + K_g)$.

The proof of this result is based on standard arguments of the closed-loop stability of delayed MIMO system [2]. It is provided in details in [4]. The remaining aims at giving some insights on how condition (21) can be checked in practice. For this, we represent the feedback system (13) by the block diagram illustrated in Figure 2. This block diagram is composed of two closed-loops, one with external input (V_s, V_g) and the other with external input (V_s, V_p) . We introduce the two following transfer functions

$$H_{sp} := \frac{H_s}{1 - H_s K_p} \quad \text{and} \quad H_{sg} := \frac{H_s}{1 - H_s K_g} \quad (22)$$

where H_{sp} and H_{sg} replace the closed-loop marked in dotted and dashed rectangle respectively and which are calculated between V_s and E_s .

Lemma 2: Suppose that H_g and the feedback systems (H_s, K_p) and (H_s, K_g) are all input-output stable. Then

$$\begin{aligned} \text{ind}(1 - H_s(K_p + K_g)) &= \text{ind}(1 - H_{sg} K_p) \\ &= \text{ind}(1 - H_{sp} K_g). \end{aligned}$$

According to Lemma 2, if the transfer function H_g and the feedback systems (H_s, K_p) and (H_s, K_g) are input-output stable, then the stability analysis of the feedback system (13) can be equivalently achieved by studying one of the two feedback systems (H_{sp}, K_g) or (H_{sg}, K_p) . In what follows, we choose to focus on the feedback system (H_{sp}, K_g) to study the stability of the network of Figure 2. This choice is motivated by the following lemma, which provides conditions under which the gain $\gamma_{H_{sp}}$ is monotonically decreasing, thus considerably simplifying the Nyquist plot analysis.

Lemma 3: Consider the transfer function H_{sp} defined by (22). For each positive δ_p as defined in (20), there exists a positive gain $c_p^*(\delta_p) > 0$ such that the loop gain $\gamma_{H_{sp}}$ is strictly decreasing for every $c_p = c_s^p c_p^s \in (0, c_p^*(\delta_p))$.

In order to prove the stability of H_{sp} , we can directly invoke [17, Theorem 2]. While the stability of H_{sp} does not depend on the loop delay δ_p , this is not the case for H_{sg} . In particular, when $c_g \sigma_g^* \sigma_s^* > 1 + c_g^g \sigma_g^*$, only a finite upper bound on the delay δ_g can be tolerated. These two observations are formalized by the following statement, proved in [4].

Lemma 4: Consider the transfer functions H_{sp} and H_{sg} , defined in (22), and let

$$G_{sp} := c_p H_s H_p \quad \text{and} \quad G_{sg} := -c_g H_s H_g, \quad (23)$$

where H_s, H_g and H_p are defined in (12). Consider the constants σ_i^* , $i \in \{s, p, g\}$, defined by (6). Then following facts hold:

- 1) The transfer function H_{sp} is input-output stable if and only if $\delta_p < \Delta(G_{sp})$. If $c_p \sigma_p^* \sigma_s^* < 1$ then $\Delta(G_{sp}) = +\infty$, otherwise $\Delta(G_{sp}) \leq 0$.
- 2) Suppose that H_g is input-output stable. Assuming that the gain $\gamma_{H_{sg}}$ is strictly decreasing, the transfer function H_{sg} is input-output stable if and only if $\delta_g < \Delta(G_{sg})$. If the inequality $c_g \sigma_g^* \sigma_s^* < 1 + c_g^g \sigma_g^*$ then $\Delta(G_{sg}) = +\infty$, otherwise $\Delta(G_{sg}) \in (0, +\infty)$.

With Proposition 3 and Lemmas 2 and 4 at hand, we are now ready to state the following result, which provides conditions for the local exponential stability of the linear system (5).

Theorem 2: Consider the delayed differential equation defined by (5). Let $u^* \in \mathbb{R}^3$ be any input such that, for the equilibrium x^* associated to these inputs, the transfer functions H_g, H_{sp} and H_{sg} , defined in (12) and (22), are input-output stable. Define $H = c_g H_g H_{sp}$. Assume that the gain of γ_H is strictly decreasing (which can be verified using Lemma 3 and [17, Lemma 4] together). For every $\delta_p > 0$, x^* is locally exponentially stable for (5) if and only if $\delta_g < \Delta(H)$.

The previous result gives only conditions for the stability of the linearized system. Nevertheless, in our case, when the linearized system is exponentially stable its hyperbolicity also implies stability properties for the original nonlinear system (1). For more details on this point, we refer the reader to [5, Theorem 4.6] and to [6, Section 10.1].

Using Lemma 3 and the fact that the transfer function H_g can be strictly decreasing [17, Lemma 4], the following result derives directly from Theorem 2.

Corollary 1: Consider the delayed differential equation defined by (5). Let $u^* \in \mathbb{R}^3$ be any input such that, for the equilibrium x^* associated to these inputs, the transfer functions H_g, H_{sp} and H_{sg} are input-output stable. Assume that the gain of γ_{H_g} is strictly decreasing. For each $\delta_p > 0$, there exist $c_p^* > 0$ such that for each $c_p < c_p^*$, x^* is locally exponentially stable for (5) if and only if $\delta_g < \Delta(H)$.

V. NUMERICAL SIMULATIONS

As in [15], the activation functions S_s, S_g and S_p are approximated by a normalized sigmoid function of the form

$$S_i(x) = \frac{B_i}{B_i + (M_i - B_i)e^{-4x}}, \quad i \in \{s, g, p\} \quad (24)$$

where B_i and M_i are given in Table I. These functions satisfy Assumption 1 with $\sigma_i = 1$, for $i \in \{s, g, p\}$. The parameter values of system (1), and precisely the transmission delays, time-constants, and activations functions, are chosen as follows. For the STN and GPe nuclei, these values are the same as those taken in [15], [17] and they are given in Table I. For the PPN, and

TABLE I

Parameter	Value	Description
δ_s^s	6 ms	Delay from STN to GPe
δ_s^g	6 ms	Delay from GPe to STN
δ_p^s	6 ms	Delay from STN to PPN
δ_s^p	6 ms	Delay from PPN to STN
δ_g^g	4 ms	Internal delay in the GPe
τ_s	6 ms	STN time constant
τ_g	14 ms	GPe time constant
τ_p	6 ms	PPN time constant
M_s	300 spk/s	Maximal firing rate for the STN
B_s	17 spk/s	Firing rate at rest for the STN
M_g	400 spk/s	Maximal firing rate for the GPe
B_g	75 spk/s	Firing rate at rest for the GPe
M_p	300 spk/s	Maximal firing rate for the PPN
B_p	17 spk/s	Firing rate at rest for the PPN

since no experimental data is available in the literature, the parameters values are taken equal to those that correspond to the STN.

As in [15], [17] the interconnection gain c_j^i from nucleus i to nucleus j ($i, j \in \{s, g\}$) in the STN-GPe network, is given by

$$c_j^i = c_j^{iH} + k(c_j^{iD} - c_j^{iH}) \quad i, j \in \{s, g\} \quad (25)$$

where k is a parameter that describes the evolution of Parkinson's disease, c_j^{iH} and c_j^{iD} are, respectively, the interconnection gains for the healthy and diseased states (given in Table II). Similarly, the external inputs are given by

$$u_i = u_i^H + k(u_i^D - u_i^H) \quad i \in \{s, g, p\} \quad (26)$$

where u_i^H and u_i^D are, respectively, the external inputs for the healthy and diseased state (given in Table II). The external inputs to the PPN, for the healthy and diseased state, are picked equal to that of STN. The parameter k is fixed to $k = 0.2$, value for which the two-dimensional system STN-GPe is locally asymptotically stable [15], [17], [18]. The evolution of system (1) is carried out in function of the interconnection gains c_p^s and c_s^p , which are taken equal.

One can easily check the dependency on the parameter c_p (defined in (20)) of the delay margin $\Delta(H)$, by plotting $\Delta(H)$ as a function of $c_p \in (0; 1)$ (see [4]). It can be observed that the delay margin decreases when c_p increases. In addition, one can see that when $c_p = 0.2$ the linear system (5) is approximately at the bifurcation point. After checking the dependency on c_p of $\Delta(H)$, we set two distinct values of c_p around the bifurcation point: $c_p = 0.1$ and $c_p = 0.3$ (note that for these values the gain $\gamma_{H_{sp}}$ is strictly decreasing). We simulate the evolution of nonlinear system (1) together with Nyquist diagram of its linearization (5) in both cases. The results of our simulations are presented in Figure 3A and 3B. When $c_p = 0.1$, we have $\Delta(H) > \delta_g$, the Nyquist plot does not encircle the critical point, and the nonlinear

TABLE II

Parameter	Healthy state	Diseased state
c_g^s	14.3	15
c_s^g	1.5	14.3
c_g^g	6.6	12.3
u_s	0.2	0.8
u_g	0.1	0.7
u_p	0.2	0.8

system (1) exhibits no oscillations after transients. When $c_p = 0.3$, we have $\Delta(H) < \delta_g$, the critical point is encircled, and the nonlinear system (1) shows sustained oscillations. These oscillations are in the beta band (28 Hz), hence likely to yield pathological symptoms.

VI. CONCLUSION

Based on a firing-rate model, we have studied the role of an external excitatory nucleus (PPN) in the generation of pathological oscillations within the basal ganglia. By Theorem 1, we show the capacity of the weight of interconnections between the PPN and the basal ganglia to change the firing-rate steady states of each nucleus. Then, by Theorem 2, we show how the transmission delays can intervene together with the interconnection gains in the modulation of these oscillations. These results are illustrated with numerical simulations.

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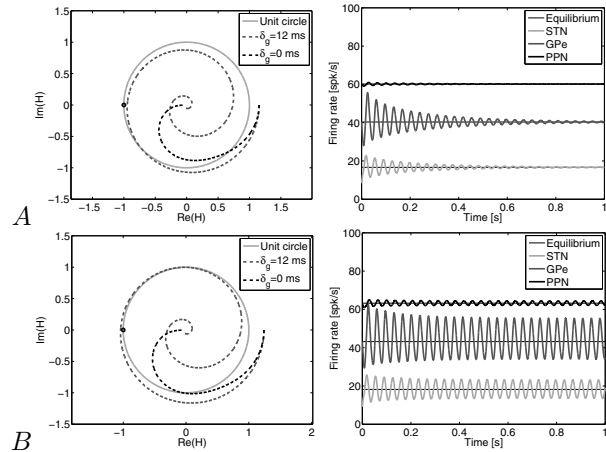


Fig. 3. Influence on stability of the interconnection gain c_p^s and c_s^p , for $k = 0.2$. On the left, the open-loop frequency-response is represented in a Nyquist diagram. On the right, the temporal evolution of the system (1) is plotted.

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